Design and Implementation of RS (255, 223) Detecting Code in FPGA

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ABSTRACT

Reed-Solomon (RS) codes are commonly used in the digital communication. It has high capability to eliminate both random errors and burst errors. In this work, the encoding of RS(255, 223) code is designed, synthesized, and simulated using Verilog language with the device family of virtex4 & device of xc4vfx12 & compare the result with device family Spartan3E & device XC3S100E. During the transfer of message, the data might get corrupted due to lots of disturbances in the communication channel. So it is necessary for the decoder tool to also have a function of correcting the error that might occur. So, from syndrome input-output waveform, it has been checked that whether there is any error in the received codeword or not. RS codes are type of burst error detecting codes which has got many applications due to its burst error detection and correction nature. This code is defined over a Galois Field $GF(2^8)$ and has the capability of correcting up to sixteen short bursts of errors.

Keywords

Reed-Solomon code, Linear Feedback Shift Register, Galois Field, Generator Polynomial, Encoder, Constant Multiplier, Syndrome, Verilog language.

1. INTRODUCTION

Reed-Solomon (RS) code which was discovered by Irving S. Reed and Gustave Solomon in Lincoln Laboratory of MIT. Massachusetts in 1960. It is a kind of multi-Bose-Chaudhuri-Hocquenghem (BCH) code with high error correction capability, which is presently one of the most effective and widely used for error control codes [1]. For the revolution of telecommunication, RS code has large contribution [2]. Specifically, RS codes can be used in computer memory and non-volatile memory applications. They are the most frequently used digital error control codes in the world [3]. The RS encoder algorithm is simpler than RS decoder and the most significant components are multipliers. Although the error correcting capability of RS codes is beyond satisfaction, because of the lack of efficient decoding algorithms they were not largely applied in their early years. W.W. Peterson firstly recognized RS codes as a special class of BCH codes [4]. Compared with other linear block codes, in the same coding efficiency, RS code has strong error correction capability and its error correction performance is close to the theoretical value, particularly on the short yards of medium. Not only RS code can correct the random error, but also it corrects unexpected error [5]. Therefore, it is widely used in deepspace communications systems, data storage systems and digital television transmission [6]. RS code is preferred in terrestrial broadcast channel, because it is a mixed channel which has both random error and burst error. In 1977, in the form of concatenated codes, RS codes were notably applied in the Voyager program [7]. In 1982, with the compact disc, there was the first commercial application in mass-produced

consumer products, where two interleaved RS codes are used [8]. Today, RS codes are largely implemented in digital storage devices and digital communication standards, though, by more modern low-density parity-check (LDPC) codes or turbo codes, they are being slowly replaced [9]. For example, RS codes are used in the Digital Video Broadcasting (DVB) standard DVB-S, but LDPC codes are used in its successor DVB-S2. RS code belongs to a family of error-correction algorithms known as BCH [10-13]. To process message data, BCH algorithms use finite fields and to detect errors in the encoded data, they use polynomial structures, called "syndromes," [14]. They can determine the presence of errors and compute the correct values by adding the check symbols to the data block. BCH algorithms have strict control over the number of check symbols [15]. Design of some other RS code like RS (204, 188) & RS (255, 251) in FPGA were performed by H. Zhang (California State University, Northridge) & A. S. Das et. al. respectively [16]. RS code is also a linear and polynomial algorithm as it processes message data as discrete blocks and it is used in modular polynomials. J. Bhaumik et. al., proposed a programmable RS encoder [17]. The received codeword is entered to RS decoder to be decoded, the decoder first tries to check if this codeword is a valid codeword or not. If it does not, errors occurred during transmission. This part of the decoder processing is called error detection, which is done by syndrome. If errors are detected, the decoder tries to correct this error using error correction part by using different algorithms [18-20].

In this work, the encoding of RS(255, 223) code is designed, synthesized and simulated using Verilog language with the help of 32 constant multipliers and by syndrome's simulation waveform, it has been checked whether the received codeword is error free or not. To get the result of encoder, firstly these multipliers are designed, synthesized and simulated using Verilog language. Before proceeding for the main program of encoder, these results are checked in Matlab code also. In the same way, the syndrome is also designed by using 32 syndrome blocks. In Section 2, RS(255, 223) encoder and syndrome are discussed briefly. Synthesis results and simulation waveforms are given in Section 3. In Section 4, performance comparison of RS(255, 223) encoder is shown. Future work is mentioned in Section 5. The paper is concluded in Section 6.

2. RS (255, 223) ENCODER

The topics, discussed in this Section are the elements of $GF(2^8)$, characteristics of RS(255, 223) code, RS encoder block diagram, the design of RS(255, 223) encoder using Linear Feedback Shift Register (LFSR) and basic idea of syndrome.

2.1 Elements of GF(2⁸)

Finite field or Galois field is an algebraic theory raised by French mathematics genius Évariste Galois. Galois fields are very important in coding theory. The RS codes studied in this paper are based on finite fields.

The elements of RS code discussed in this paper are on the field GF(2⁸) = 256. There are 2⁸ = 256 elements on GF(2⁸), among which 255 elements are non-zero [14]. The primitive polynomial on GF (2⁸) is $p(x) = x^8 + x^4 + x^3 + x^2 + 1$. From the primitive polynomial $p(\alpha) = x^8 + x^4 + x^3 + x^2 + 1 = 0$, the elements with order greater than "7" can be derived. The 256 elements on field GF (2⁸) are shown in Table 1.

 Table 1: Elements of Field GF(2⁸)

Power $(\alpha)^i$	Polynomial Form	Binary Form	Decimal Form
0	0	00000000	0
α ⁰	1	00000001	1
α^1	α	00000010	2
α^2	α^2	00000100	4
α ³	α ³	00001000	8
α^4	α4	00010000	16
α^5	α ⁵	00100000	32
α^6	α ⁶	01000000	64
α^7	α ⁷	10000000	128
α ⁸	$\begin{array}{c} \alpha^4 + \alpha^3 \\ + \alpha^2 + 1 \end{array}$	00011101	29
α9	$\alpha^5 + \alpha^4 + \alpha^3 + \alpha$	00111010	58
α^{10}	$\alpha^6 + \alpha^5 + \alpha^4 + \alpha^2$	01110100	116
α^{11}	$\alpha^7 + \alpha^6 + \alpha^5 + \alpha^3$	11101000	232
α^{12}	$\frac{\alpha^7 + \alpha^6 +}{\alpha^3 + \alpha^2 + 1}$	11001101	205
α ¹³	$\frac{\alpha^7 + \alpha^2 + \alpha}{+ 1}$	10000111	135
α ¹⁴	$\alpha^4 + \alpha + 1$	00010011	19
α^{253}	$\frac{\alpha^6 + \alpha^2}{+ \alpha^1 + 1}$	01000111	71
α^{254}	$\begin{array}{c} \alpha^7 + \alpha^3 \\ + \alpha^2 + \alpha^1 \end{array}$	10001110	142

2.2 Characteristics of RS(255, 223) code

The characteristics of RS(255, 223) code are discussed in this paper are as below:

Degree of the Polynomial: m = 8Code Length: n = 255Information Symbols: k = 223 Parity Check Symbols: r = n - k = 2t = 32Minimum Distance: dmin = n - k + 1 = 2t + 1 = 33Error Correcting Capability: t = 16Code Rate = Code Efficiency = k/n = 223/255 = 0.875

However, each symbol is represented by eight binary digits or one byte. Also, each data block contains 223 information symbols. This code is capable of correcting up to sixteen short burst errors of one byte or any burst error combination of up to a total length of eight bytes, providing that they only affect a maximum of sixteen individual symbols [15].

2.3 Construction of GF(2⁸)

The elements of $GF(2^8)$ are generated by primitive polynomial of degree 8.

$$p(x) = X^8 + X^4 + X^3 + X^2 + 1$$

Let α be the primitive element in GF(2⁸) and the root of p(X) Then.

$$p(x) = X^8 + X^4 + X^3 + X^2 + 1 = 0$$

Or
$$X^8 = X^4 + X^3 + X^2 + 1$$

So, the elements can be represented in an 8-tuple with 8 components being 0 or 1 and represent code word [17]. The zero element of $GF(2^8)$ appears as an all zero 8-tuple.

Also, if α is a primitive element in GF(2^{*m*}), then the root of p(x) is only the first thirty-two powers of α and are the roots of the generator polynomial. Meanwhile, the generator polynomial for (255, 223) code is given by:

$$\begin{array}{l} g(x) = & (x + \alpha)(x + \alpha^2)(x + \alpha^3)(x + \alpha^4)(x + \alpha^5)(x + \alpha^6) \\ & (x + \alpha^7)(x + \alpha^8)(x + \alpha^9)(x + \alpha^{10})(x + \alpha^{11})(x + \alpha^{12}) \\ & (x + \alpha^{13})(x + \alpha^{14})(x + \alpha^{15})(x + \alpha^{16})(x + \alpha^{17})(x + \alpha^{18}) \\ & (x + \alpha^{19})(x + \alpha^{20})(x + \alpha^{21})(x + \alpha^{22})(x + \alpha^{23})(x + \alpha^{24}) \\ & (x + \alpha^{25})(x + \alpha^{26})(x + \alpha^{27})(x + \alpha^{28})(x + \alpha^{29})(x + \alpha^{30}) \\ & & (x + \alpha^{31})(x + \alpha^{32}) \end{array}$$

Therefore, the coefficients of g(x) used in the encoder multiplication are:

 $\begin{array}{l} g_0=45,\ g_1=216,\ g_2=239,\ g_3=24,\ g_4=253,\ g_5=104,\\ g_6=27,\ g_7=40,\ g_8=107,\ g_9=50,\ g_{10}=163,\ g_{11}=210,\\ g_{12}=227,\ g_{13}=134,\ g_{14}=224,\ g_{15}=158,\ g_{16}=119,\\ g_{17}=13,\ g_{18}=158,\ g_{19}=1,\ g_{20}=238,\ g_{21}=164,\ g_{22}=82,\\ g_{23}=43,\ g_{24}=15,\ g_{25}=232,\ g_{26}=246,\ g_{27}=142,\ g_{28}=50,\\ g_{29}=189,\ g_{30}=29,\ g_{31}=232,\ g_{32}=1\end{array}$

2.4 Encoder Architecture

The block diagram of RS encoder is shown in Figure 1 [18].



Figure 1: RS Encoder Block Diagram

Encoders are designed as feedback shift register, Figure 2. The data massage blocks of 223 symbols shift sequentially as an input to the encoder and when the last message symbol is loaded, the feedback register contains the thirty-two parity check symbols. These symbols will then be shifted out following the 223 information symbols to generate a code word of 255 symbols as an output of the encoder [20].



Figure 2: RS(255, 223) Encoder using LFSR

2.5 Syndrome

Syndrome is utilized to determine whether an error happened in the transmission. If value of the syndrome is 0, there is no error in the transmission and the received sequence is the code word; while if syndrome value is non zero, then there is error, hence error correction is needed. Syndrome values are only dependent on the error pattern. It checks if there is any error in the received codeword or not.

 $\mathbf{r}(\mathbf{X}) = \mathbf{c}(\mathbf{X}) + \mathbf{e}(\mathbf{X})$

Where, code word = c(x), error pattern = e(x), received signal = r(x)

There are 32 syndrome blocks which has helped us to get the input and output waveform of syndrome. In Table 2, the values of 32 syndrome blocks are shown. Firstly these 32 syndrome blocks are synthesized and simulated using Verilog

code with the device family of virtex4 & device of xc4vfx12. The syndrome blocks' simulation result is checked by Matlab code. Then the syndrome is synthesized and simulated with the help of these syndrome blocks.

SYNDRO	ME BLOCKS
$S_1 = 2$	$S_{17} = 152$
$S_2 = 4$	$S_{18} = 45$
$S_3 = 8$	$S_{19} = 90$
$S_4 = 16$	$S_{20} = 180$
$S_{5} = 32$	$S_{21} = 117$
$S_{6} = 64$	$S_{22} = 234$
$S_7 = 128$	$S_{23} = 201$
S ₈ = 29	$S_{24} = 143$
$S_9 = 58$	S ₂₅ = 3
$S_{10} = 116$	S ₂₆ = 6
$S_{11} = 232$	$S_{27} = 12$
$S_{12} = 205$	$S_{28} = 24$
$S_{13} = 135$	$S_{29} = 48$
$S_{14} = 19$	$S_{30} = 96$
$S_{15} = 38$	S ₃₁ = 192
$S_{16} = 76$	S ₃₂ = 157

3. SYNTHESIS RESULTS AND SIMULATION WAVEFORMS

In this section, synthesis results and simulation waveforms of 32 coefficients, RS(255, 223) encoder, 32 syndrome blocks and syndrome have been elaborated.

In Table 3, synthesis result of 32 coefficients used in RS(255, 223) encoder is shown. These results are got in Verilog language using device family virtex4 and device xc4vfx12.

 Table 3: Synthesis Result for Coefficients of Generator

 Polynomial g(x)

Coefficients	Number of Slices	Number of 4 Input LUTs	Number of Bounded IOBs	Delay (ns)
1			16	3.562
13	6	10	16	4.95
15	7	12	16	4.97
24	5	8	16	4.288
27	6	10	16	4.949
29	5	9	16	4.949
40	6	10	16	4.957
43	5	8	16	4.281
45	6	10	16	4.957
50	6	10	16	4.949
104	5	9	16	4.971
107	8	14	16	4.949

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119	6	11	16	4.971
134	4	7	16	4.283
142	2	3	16	4.281
158	4	7	16	4.288
163	7	13	16	4.957
164	5	8	16	4.282
189	6	11	16	4.963
210	7	12	16	4.961
216	3	6	16	4.288
224	6 6	11	16	4.971
227		11	16	4.971
232	6	10	16	4.957
238	5	9	16	4.971
239	7	12	16	4.957
246	8	14	16	4.971
253	7	13	16	4.895

After getting these synthesis result, all the 32 coefficients are simulated. The simulation waveform in Figure 3, for coefficient g_0 is shown. All the results are checked using Matlab code.





Using Matlab code, we can Verify the Simulation Result for Coefficient $g_0 = 45$:

This code is defined over Galois Field $GF(2^8)$ and the primitive polynomial is $X^8 + X^4 + X^3 + X^2 + 1$ (285 decimal). For the coefficient $g_0 = 45$, if we multiply it by 1, 2, 4, 8, 16, 32, 64, 128. The result will come 45, 90, 180, 117, 234, 201, 143, 3.

In Table 4, synthesis result of RS(255, 223) encoder is shown. These results are got in Verilog language using device family virtex4 and device xc4vfx12.

Table 4: Synthesis Result of RS(255, 223) Encoder

Numb er of Slices	Numb er of Slice Flip Flops	Numb er of 4 Input LUTS	Numb er of IOS	Numb er of Boun ded IOBS	Numb er of GCL KS	Delay (ns)
239	256	455	22	22	1	3.014

After getting the synthesis result, the input and output waveform for RS(255, 223) encoder is shown in Figure 4 for the input 1 in all 223 input message signal.







In Table 5, synthesis result of syndrome blocks is shown. These results are got in Verilog language using device family virtex4 and device xc4vfx12.

Syndrome Blocks	Number of Slices (out of 5472)	Number of 4 Input LUTs (out of 10944)	Delay (ns)
$S_1 = 2$	2	3	4.868
S ₂ = 4	2	4	4.946
S ₃ = 8	3	5	4.986
S ₄ = 16	3	6	4.986
S ₅ = 32	4	7	4.986
S ₆ = 64	4	8	4.993
S ₇ = 128	4	8	4.993
S₈ = 29	5	9	4.949
S ₉ = 58	5	9	5.277
S₁₀ = 116	5	9	5.277
S ₁₁ = 232	6	10	4.957
S₁₂ = 205	5	9	5.582
S₁₃ = 135	5	9	5.284
S₁₄ = 19	5	9	5.419
S ₁₅ = 38	5	9	5.541
S ₁₆ = 76	5	9	5.569
S ₁₇ = 152	5	9	5.627
S ₁₈ = 45	6	10	4.957
S ₁₉ = 90	5	9	5.284
$S_{20} = 180$	4	8	5.543
S₂₁ = 117	5	9	5.294
S₂₂ = 234	4	8	5.542
S₂₃ = 201	4	8	4.986
S₂₄ = 143	4	8	4.953
S ₂₅ = 3	4	8	4.949
$S_{26} = 6$	4	8	4.989
S ₂₇ = 12	4	8	4.993
S₂₈ = 24	5	8	4.288
S ₂₉ = 48	4	8	5.543
S ₃₀ = 96	4	8	5.543
S₃₁ = 192	4	8	5.620
S ₃₂ = 157	4	8	5.660

Table 5: Synthesis Result of Syndrome Blocks

After getting the synthesis result, the input and output waveform for syndrome block S_3 is shown in Figure 5. All the syndrome blocks are simulated using Verilog language and the results are checked by Matlab code.



Figure 5: Simulation Waveform of Syndrome Block S₃= 8

Using Matlab code, we can Verify the Simulation Result for Coefficient $S_3 = 8$:

This code is defined over Galois Field $GF(2^8)$ and the primitive polynomial is $X^8 + X^4 + X^3 + X^2 + 1$ (285 decimal). For the coefficient $S_3 = 8$, if we multiply it by 1, 2, 4, 8, 16, 32, 64, 128. The result will come 8, 16, 32, 64, 128, 29, 58, 116.

In Table 6, synthesis result of syndrome is shown. The result is got in Verilog language using device family virtex4 and device xc4vfx12.

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Number of Slices	Number of Slice Flip Flops	Number of 4 Input LUTS	Number of GCLKS	Delay (ns)
408	256	789	1	2.131

After getting the synthesis result, the input and output waveform for syndrome is shown in Figure 6.

In the input of this waveform it is seen that, 32 parity bits are given, which are got from simulation result of RS(255, 223) Encoder in Figure 4, when input is '1' for all 223 input message signal. In the simulation result, syndrome is non zero, so there is error in the received codeword, hence error correction is needed.

Current Simulation Time: 301000 ns			1	19000		1		49500	1	50	000	T	50500		
0 (1 10[7:0]	67	163	93	X	54	216		183	251	19	227	234	(161)	67	
D 🛃 y11[7:0]	102	33	76	X	36	213	X	210	162	191	17	(183)	81	102	
D B (y12[7 0]	15	186	71	X	12	14	X	78	61	162	106	(30)	(103)	15	
D (1 1)13[7 0]	124	134	6	X	86	173	X	223	19	226	93	104	240	124	
D (H y14[7 0]	205	199	(116	X	63	(111	X	41	181	130	70	151	(196)	205	
D (1 415[7 0]	158	139	94	XI	41	172	X	252	139	189	211	117	204	158	
■ 84 y20[7:0]	166	139	(17	χ	74	55	X	100	221	222	91	41	(122)(166	
B (1 y16[7 0]	232	228	239	X	103	25	X	110	254	224	76	(17)	107	232	
D (1 y21[7.0]	222	50	218	X	16	(115	X	6	88	231	104	33	96	222	
D (%1 /1717 0)	61	223	201	X	32	110	X	234	116	228	255	47	249	61	1
a (1 y22[7.0]	39	92	32	X	36	69	X	205	130	21	38	162	(113)	39	
0 🕅 y18[7 0]	178	122	2	XI	36	178	X	195	245	150	238	(151)	210	178	
D B1 y23[7:0]	48	199	(51	XI	52	138	X	248	154	45	85	199	172	48	
0 81 y19[7 0]	83	196	249	XI	99	(12)	X	95	(117	48	181	232	70	83	
B 🛃 y24[7 0]	65	91	244	χ	61	25	X	28	58	180	33	85	(121)	85	
0 (%) y25[7 0]	92	186	(162	X	115	150	X	100	91	61	(110	79	57	92	
0 84 y30[7 0]	192	6	(163	X	08	28	X	194	93	133	147	99	(180)(192	
D (\$1 y26[7 0]	107	151	179	χ	59	119	X	105	200	243	192	95	37	107	
0 (1 y31[7 0]	22	92	122	XI	24	(119	X	56	237	94	164	(1)	(114)	22	
0 81 y27[7 0]	166	202	59	XI	10	167	X	196	172	225	13	(142)	(138)	166	
■ 🛃 y28[7.0]	117	107	6	XI	56	1	X	1	142	228	119	83	(109)	117	
D \$1 y29[7:0]	248	46	255	X	55) 22	X	204	203	18	110	245	(198)	248	
D (1 ¥0[7 0]	15	153	54	X	80	254	X	214	175	185	143	53	236	15	-
	15	153	54	X	80	25	X	214	175	185	143	53	236	15	

Current Simulation Time: 301000 ns		ï	49(000 		49500		50	000 		50500	
■ <mark>81</mark> y31[7:0]	22	92	122	124	119	58	237	94	164		114	22
🖬 🚰 y27[7:0]	166	202	59	110	187	198	172	225	13	142	138	166
■ <mark>61</mark> y28[7:0]	117	107	6	156		1	142	226	119	83	109	117
🖬 😽 y29[7:0]	248	46	255	155	221	204	203	18	110	245	198	248
🖬 😽 y0[7:0]	15	153	- 54	80	254	214	175	185	143	- 53	236	15
🖬 🔂 y1[7:0]	68	33	157	114	139	33	154	168	122	195	173	.68
🖬 😽 y2[7:0]	218	245	226	127	129	75	124	61	21	158	2	218
■ <mark>61</mark> ¥3[7:0]	60	173	21	145	187	72	234	252	155	115	229	60
🖬 😽 y4[7:0]	71	130	148	75	203	111	127	181	121	173	130	71
🖬 😽 y5(7:0)	66	54	24	114	143	- 87	122	17	212	21	175	66
🖬 😽 y6[7:0]	68	- 58	\$3	157	107	156	171	197	188	141	91	68
🖬 😽 y7(7:0)	239	-24	60	234	62	219	41	136	46	23	24	239
🖬 😽 y8[7:0]	211	1	25	76	93	137	165	119	244	32	149	211
■ <mark>61</mark> y9[7:0]	90	180	121	182	214	213	8	144	107	71	155	90
PERIOD[31:0]	200						2	10				
OUTY_CYCLE	0.5						0	5				
OFFSET[31:0]	100						1))				
o cik	1											
o enable	0									_	_	
o int	0											
🖬 😽 u(7:0)	0	41	25	80	94	55	30	250	224	54	134	202 X 0
o, shift	0											
o cim	1											

Figure 6: Simulation Waveform of Syndrome

4. PERFORMANCE COMPARISON

In Table 4, the synthesis result of RS(255, 223) encoder by using the device family of Virtex4 & device of Xc4vfx12 is shown. It is compared with device family of Spartan3E & device of XC3S100E. The performance comparison between these two is listed in Table 4, where it is seen that number of Slice Flip-Flops, IOBS, bounded IOBS & GCLKS are same for both the cases. But for device family Virtex4, number of Slices is needed more than device family Spartan3E, whereas Spartan3E's delay is greater than the device family Virtex4.

Table 7: Performance Comparison of Synthesis Result for RS(255, 223) encoder

Device Used	Nu mb er of Slic es	Num ber of Slice Flip Flop s	Nu mb er of 4 Inp ut LU TS	Nu mb er of IOS	Num ber of Boun ded IOB S	Nu mb er of GC LK S	Dela y (ns)
Virtex4 & device: Xc4vfx 12	239	256	455	22	22	1	3.014
Spartan 3E & device: XC3S1 00E	238	256	453	22	22	1	6.124

5. FUTURE WORK

In this work, the synthesis result with simulation waveform of RS(255, 223) encoder and syndrome are shown. During the transfer of message, the data might get corrupted due to lots of disturbances in the communication channel. In the syndrome's input-output waveform, it is seen that the value of the syndrome is non zero. So there is error in the transmitted codeword. By correcting these errors, we can recover the actual codeword. So the future work will be,

- A. Determine the roots of which are related to the error locations using Chien Search
- **B.** Calculate the values of the error evaluator using Forney Algorithm
- C. Recover the corrected codeword by adding E(x) with R(x)

6. CONCLUSION

The communication channel in modern digital and data storage systems requires error detecting and correcting codes to correct the errors that occur during the transmission of data. It can be also implemented on Visual Sensor Network (VSN), Deep-Space communication and Digital μ -wave radio. In this paper, RS encoding, system specification of RS (255, 223) encoder with its architecture & design using LFSR are discussed. The co-efficients of generator polynomial used in the encoder multiplication are mentioned. These terms are simulated using Verilog code & whether the simulation results are right or wrong, are tested by Matlab code which has helped to design encoder. With the help of these co-efficient

terms, the simulation waveform of RS(255, 223) encoder is got by using the Verilog code and performance comparison of RS(255, 223) encoder using a different device is shown. Lastly syndrome is simulated by using the syndrome blocks which is also shown in this paper. As the received codeword is erroneous, so error correction is necessary. So, in this paper, it is detected whether the received code word is error free or not.

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8. REFERENCES

- I. S. Reed and G. Solomon. Polynomial codes over certain finite fields. Journal of the Society for Industrial and Applied Mathematics, 8:300-304, 1960.
- [2] ETSI, 1997. Digital broadcasting systems for television, sound and data services; Framing structure, channel coding and modulation for digital terrestrial television. European Telecommunication Standard ETS 300 744.
- [3] S.B. Wicker. Error Control Systems for Digital Communication and Storage. Prentice-Hall, Englewood Clips, NJ, 1995.
- [4] G. D. Forney, Jr. On decoding BCH codes. IEEE Transactions on Information Theory, IT-11:549-557, 1965.
- [5] S. Liu and Jr. D. J. Costello. Error Control Coding: Fundamentals and Applications. Prentice-Hall, Englewood Clips, NJ, 1983.
- [6] Akyildiz, I.F., T. Melodia, and K.R. Chowdury, Wireless multimedia sensor networks: A survey. IEEE Wireless Communications. 2007, 14(6): p. 32-39.
- [7] H. Y. Hsu and A. Y. Wu, "VLSI Design of a Reconfigurable Multimode Reed-Solomon Codec for High Speed Communication Systems," in IEEE Asia-Pacific Conference on ASIC, 2002, pp. 359-362.
- [8] K.A.S. Immink, \Reed Solomon codes and the compact disc," in Reed-Solomon Codes and Their Applications, eds. S.B. Wicker and V.K. Bhargava. New York: IEEE Press, 1994, pp. 41-59.
- [9] Stephen B. Wicker and Vijay K. Bhargava, "Reed-Solomon codes and their applications", IEEE Press, New Jersey, 1994.

- [10] J. L. Massey. Shift register synthesis and BCH decoding. IEEE Trans-actions on Information Theory, pages 122 -127, January 1969.
- [11] Sklar B, "Digital Communication: Fundamentals and Applications", Second Edition, Prentice-Hal, 2001.
- [12] Pretzel, O. 1992.Error-correcting codes and finite fields. Clarendon Press, Oxford, 1992.
- [13] R. T. Chien. Cyclic decoding procedures for Bose-Chaudhuri-Hocquenghem codes. IEEE Transactions on Information Theory, IT-10:357 - 363, October 1964.
- [14] Y. R. Shayan and T. Le-Ngoc, "Decoding Reed-Solomon codes generated by any generator polynomial," Electronics Letters, vol. 25, no. 18, Aug. 1989, pp. 1223-1224.
- [15] T. Yaghoobian and I.F. Blake "Reed Solomon and Algebraic Geometry Codes," in Reed-Solomon Codes and Their Applications, eds. S.B. Wicker and V.K. Bhargava. New York: IEEE Press, 1994, pp. 292-314.
- [16] Anindya Sundar Das, Satyajit Das and Jaydeb Bhaumik, Design of RS (255, 251) Encoder and Decoder in FPGA. International Journal of Soft Computing and Engineering (IJSCE) ISSN: 2231-2307, Volume-2, Issue-6, January.
- [17] Jaydeb Bhaumik, Anindya Sundar Das and Jagannath Samanta, Architecture for Programmable Generator Polynomial Based Reed-Solomon Encoder and Decoder. International Journal of Soft Computing and Engineering (IJSCE) ISSN: 2231-2307, Volume-2, Issue-6, January
- [18] J. Bhaumik and D. Roy Chowdhury, "An Integrated ECC-MAC Based on RS Code," Transactions on Computational Science, vol. IV, LNCS5430, Apr. 2009, pp. 117-135.
- [19] Syed Shahzad Sha, Saqib Yaqub, and Faisal Suleman, " Self-correcting codes conquer noise Part 2: Reed-Solomon codec's", EDN, pp. 107-120, March 2001.
- [20] J. Jittawutipoka and J. Ngarmnil, "Low complexity Reed-Solomon encoder using globally optimized finite field multipliers", TENCON 2004, 2004 IEEE Region 10 Conference, vol. D, 21-24 Nov. 2004, pp. 423-426, doi: 10.1109/TENCON.2004.1414960.